

FW803 PHY *IEEE*® 1394A Three-Cable Transceiver/Arbiter Device



Distinguishing Features

- Supports extended BIAS_HANDSHAKE time for enhanced interoperability with camcorders.
- While unpowered and connected to the bus, will not drive TPBIAS on a connected port even if receiving incoming bias voltage on that port.
- Does not require external filter capacitors for PLL.
- Does not require a separate 5 V supply for 5 V link controller interoperability.
- Interoperable across 1394[™] cable with 1394 physical layers (PHY) using 5 V supplies.
- Interoperable with 1394 link-layer controllers using 5 V supplies.
- Powerdown features to conserve energy in batterypowered applications include:
 - Device powerdown pin.
 - Link interface disable using LPS.
 - Inactive ports power down.
- Interface to link-layer controller supports Annex lelectrical isolation as well as bus-keeper isolation.

Features

- Provides three fully compliant cable ports at 100 Mbits/s, 200 Mbits/s, and 400 Mbits/s.
- Fully supports the 1394 Open HCI requirements.
- Supports arbitrated short bus reset to improve utilization of the bus.
- Supports ack-accelerated arbitration and fly-by concatenation.
- Supports connection debounce.
- Supports multispeed packet concatenation.
- Supports PHY pinging and remote PHY access packets
- Fully supports suspend/resume.

- Supports PHY-link interface initialization and reset.
- Supports the 1394a-2000 register set.
- Supports LPS/link-on as a part of PHY-link interface.
- Compliant with the IEEE 1394-1995 Standard for a High Performance Serial Bus and IEEE P1394a draft 2.0.
- Fully interoperable with the *FireWire*® implementation of *IEEE 1394*-1995.
- Reports cable power fail interrupt when voltage at CPS pin falls below 7.5 V.
- Provides separate cable bias and driver termination voltage supply for each port.

Other Features

- 64-pin TQFP package.
- Single 3.3 V supply operation.
- Data interface to link-layer controller provided through 2/4/8 parallel lines at 50 Mbits/s.
- 25 MHz crystal oscillator and PLL provide a 50 MHz link-layer controller clock as well as transmit/receive data at 100 Mbits/s, 200 Mbits/s, and 400 Mbits/s.
- Node power-class information signaling for system power management.
- Multiple separate package signals provided for analog and digital supplies and grounds.

Description

The Agere Systems Inc. FW803 device provides the analog physical layer functions needed to implement a three-port node in a cable-based *IEEE 1394*-1995 and *IEEE 1394*a-2000 network.

Each cable port incorporates two differential line transceivers. The transceivers include circuitry to monitor the line conditions as needed for determining connection status, for initialization and arbitration, and for packet reception and transmission. The PHY is designed to interface with a link-layer controller (LLC).

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Description (continued)

The PHY requires either an external 24.576 MHz crystal or crystal oscillator. The internal oscillator drives an internal phase-locked loop (PLL), which generates the required 393.216 MHz reference signal. The 393.216 MHz reference signal is internally divided to provide the 49.152 MHz, 98.304 MHz, and 196.608 MHz clock signals that control transmission of the outbound encoded strobe and data information. The 49.152 MHz clock signal is also supplied to the associated LLC for synchronization of the two chips and is used for resynchronization of the received data. The powerdown function, when enabled by the PD signal high, stops operation of the PLL and disables all circuitry except the cable-not-active (CNA) signal circuitry.

The PHY supports an isolation barrier between itself and its LLC. When /ISO is tied high, the link interface outputs behave normally. When /ISO is tied low, internal differentiating logic is enabled, and the outputs become short pulses, which can be coupled through a capacitor or transformer as described in the IEEE 1394-1995 Annex J. To operate with bus-keeper isolation, the /ISO pin of the FW803 must be tied high.

Data bits to be transmitted through the cable ports are received from the LLC on two, four, or eight data lines (D[0:7]), and are latched internally in the PHY in synchronization with the 49.152 MHz system clock. These bits are combined serially, encoded, and transmitted at 98.304 Mbits/s, 196.608 Mbits/s, or 393.216 Mbits/s as the outbound data-strobe information stream. During transmission, the encoded data information is transmitted differentially on the TPA and TPB cable pair(s).

During packet reception, the TPA and TPB transmitters of the receiving cable port are disabled, and the receivers for that port are enabled. The encoded data information is received on the TPA and TPB cable pair. The received data-strobe information is decoded to recover the receive clock signal and the serial data bits. The serial data bits are split into two (for S100), four (for S200), or eight (for S400) parallel streams, resynchronized to the local system clock, and sent to the associated LLC. The received data is also transmitted (repeated) out of the other active (connected) cable ports.

Both the TPA and TPB cable interfaces incorporate differential comparators to monitor the line states during initialization and arbitration. The outputs of these comparators are used by the internal logic to determine the arbitration status. The TPA channel monitors the incoming cable common-mode voltage. The value of this common-mode voltage is used during arbitration to set the speed of the next packet transmission. In addition, the TPB channel monitors the incoming cable common-mode voltage for the presence of the remotely supplied twisted-pair bias voltage. This monitor is called bias-detect.

The TPBIAS circuit monitors the value of incoming TPA pair common-mode voltage when local TPBIAS is inactive. Because this circuit has an internal current source and the connected node has a current sink, the monitored value indicates the cable connection status. This monitor is called connect-detect.

Both the TPB bias-detect monitor and TPBIAS connect-detect monitor are used in suspend/resume signaling and cable connection detection.

The PHY provides a 1.86 V nominal bias voltage for driver load termination. This bias voltage, when seen through a cable by a remote receiver, indicates the presence of an active connection. The value of this bias voltage has been chosen to allow interoperability between transceiver chips operating from 5 V or 3 V nominal supplies. This bias voltage source should be stabilized by using an external filter capacitor of approximately 0.33 μF.

The transmitter circuitry, the receiver circuitry, and the twisted-pair bias voltage circuity are all disabled with a powerdown condition. The powerdown condition occurs when the PD input is high. The port transmitter circuitry, the receiver circuitry, and the TPBIAS output are also disabled when the port is disabled, suspended, or disconnected.

The line drivers in the PHY operate in a high-impedance current mode and are designed to work with external 112 Ω line-termination resistor networks. One network is provided at each end of each twisted-pair cable. Each network is composed of a pair of series-connected 56 Ω resistors. The midpoint of the pair of resistors that is directly connected to the twisted-pair A (TPA) signals is connected to the TPBIAS voltage signal. The midpoint of the pair of resistors that is directly connected to the twisted-pair B (TPB) signals is coupled to ground through a parallel RC network with recommended resistor and capacitor values of 5 k Ω and 220 pF, respectively.

Description (continued)

The value of the external resistors are specified to meet the *IEEE 1394* standard specifications when connected in parallel with the internal receiver circuits.

The driver output current, along with other internal operating currents, is set by an external resistor. This resistor is connected between the R0 and R1 signals and has a value of 2.49 k Ω ± 1%.

The FW803 supports suspend/resume as defined in the *IEEE 1394*a-2000 specification. The suspend mechanism allows an FW803 port to be put into a suspended state. In this state, a port is unable to transmit or receive data packets, however, it remains capable of detecting connection status changes and detecting incoming TPBias. When all ports of the FW803 are suspended, all circuits except the bias voltage reference generator, and bias detection circuits are powered down, resulting in significant power savings. The use of suspend/resume is recommended.

Four signals are used as inputs to set four configuration status bits in the self-identification (self-ID) packet. These signals are hardwired high or low as a function of the equipment design. PC[0:2] are the three signals that indicate either the need for power from the cable or the ability to supply power to the cable. The fourth signal, C/LKON, as an input, indicates whether a node is a contender for bus manager. When the C/LKON signal is asserted, it means the node is a contender for bus manager. When the signal is not asserted, it means that the node is not a contender. The C (contender) bit corresponds to the C field (bit 20) in the self-ID packet. PC [0:2] corresponds to the power field of the selfID packet in the following manner: PC0 corresponds to bit 21, PC1 corresponds to bit 22, and PC2 corresponds to bit 23 (see Table 4-29 of the IEEE 1394-1995 standard for additional details). As an example, for a Power_Class value of 001, PC0 = 0, PC1 = 0 and PC2 = 1.

A powerdown signal (PD) is provided to allow a powerdown mode where most of the PHY circuits are powered down to conserve energy in battery-powered applications. The internal logic in FW803 is reset as long as the powerdown signal is asserted. A cable status signal, CNA provides a high output when none of the twisted-pair cable ports are receiving incoming bias voltage. This output is not debounced. The CNA output can be used to determine when to power the PHY down or up. In the powerdown mode, all circuitry is disabled except the CNA circuitry. It should be noted

that when the device is powered down, it does not act in a repeater mode.

When the power supply of the PHY is removed while the twisted-pair cables are connected, the PHY transmitter and receiver circuitry has been designed to present a high impedance to the cable in order to not load the TPBIAS signal voltage on the other end of the cable.

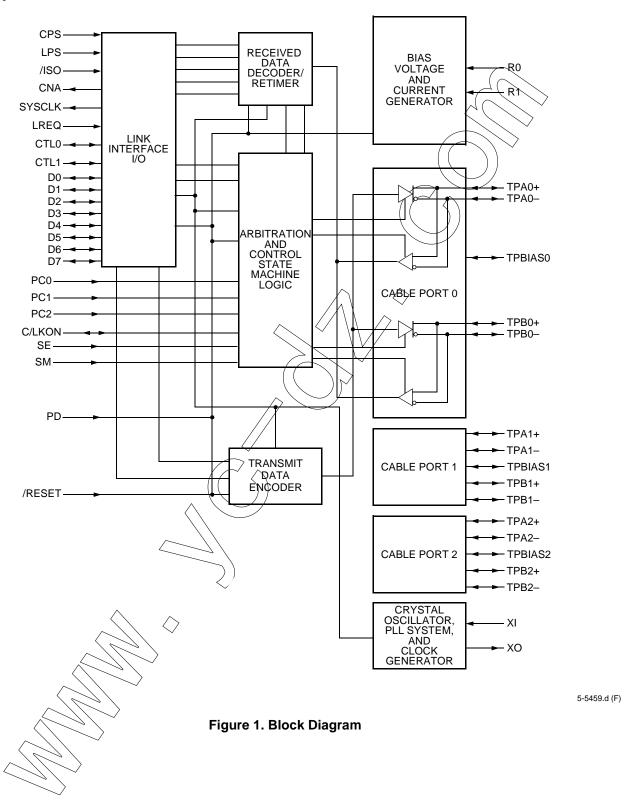
Whenever the TPA±/TPB± signals are wired to a connector, they must be terminated using the normal termination network (see Figure 4). This is required for reliable operation. For those applications, when one or more of the FW803's ports are not wired to a connector, those unused ports may be left unconnected without normal termination. When a port does not have a cable connected, internal connected effect circuitry will keep the port in a disconnected state.

Note: All gap counts on all nodes of a 1394 bus must be identical. The software accomplishes this by issuing PHY configuration packets (see Section 4.3.4.3 of the IEEE 1394-1995 standard) or by issuing two bus resets, which resets the gap counts to the maximum level (3Fh).

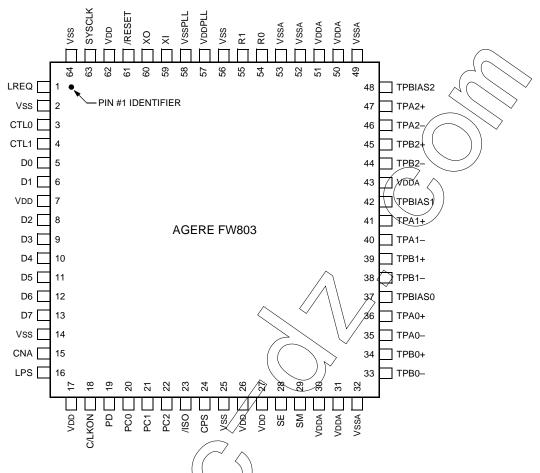
The link power status (LPS) signal works with the \mathcal{L}/\mathcal{L} KON signal to manage the LLC power usage of the node. The LPS signal indicates that the LLC of the node is powered up or powered down. If LPS is inactive for more than 1.2 us and less than 25 us, the PHY/link interface is reset. If LPS is inactive for greater than 25 µs, the PHY will disable the PHY/link interface to save power. FW803 continues its repeater function even when the PHY/link interface is disabled. If the PHY then receives a link-on packet, the C/LKON signal is activated to output a 6.114 MHz signal, which can be used by the LLC to power itself up. Once the LLC is powered up, the LPS signal communicates this to the PHY and the PHY/link interface is enabled. The C/LKON signal is turned off when LPS is active or when a bus reset occurs, provided the interrupt that caused C/LKON is not present.

Two of the FW803's signals are used to set up various test conditions used only during the device manufacturing process. These signals (SE and SM) should be connected to Vss for normal operation.

Description (continued)



Signal Information



Note: Active-low signals are indicated by "/" at the beginning of signal names, within this document.

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Figure 2. Pin Assignments

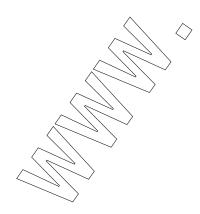


Table 1. Signal Descriptions

Pin	Signal*	Туре	Name/Description
18	C/LKON	I/O	Bus Manager Capable Input and Link-On Output. On hardware reset (/RESET), this pin is used to set the default value of the contender status indicated during self-ID. The bit value programming is done by tying the signal through a 10 k Ω resistor to VDD (high, bus manager capable) or to GND (low, not bus manager capable). Using either the pull-up or pull-down resistor allows the link-on output to override the input value when necessary.
			After hardware reset, this pin is set as an output. If the LPS is inactive, C/LKON indicates one of the following events by asserting a 6.114 MHz signal.
			 FW803 receives a link-on packet addressed to this node. Port_event register bit is 1. Any of the Timeout, Pwr_fail, or Loop register bits are 1 and the Resume_int register bit is also 1. Once activated, the C/LKON output will continue active until the LPS becomes active. The PHY also deasserts the C/LKON output when a 1394 bus reset occurs, if the C/LKON is active due solely to the reception of a link-on packet.
			Note: If an interrupt condition exists which would otherwise cause the C/LKON output to be activated if the LPS were inactive, the C/LKON output will be activated when the LPS subsequently becomes inactive.
15	CNA	0	Cable-Not-Active Output. CNA is asserted high when none of the PHY ports are receiving an incoming bias voltage. This circuit remains active during the powerdown mode.
24	CPS		Cable Power Status. CPS is normally connected to the cable power through a 400 k Ω resistor. This circuit drives an internal comparator that detects the presence of cable power. This information is maintained in one internal register and is available to the LLC by way of a register read (see Table 8, address register 00002, bit 7/PS). In applications that do not sink or source 1394 power (VP), this pin can be tied to ground.
			Note: When this pin is grounded, the Pwr_fail bit in PHY register 01012 will be set.
3	CTL0 C7L1	1 0	Control I/O. The CTLn signals are bidirectional communications control signals between the PHY and the LLC. These signals control the passage of information between the two devices. Bus-keeper circuitry is built into
5, 6, 8, 9, 10, 11,	D[0:7]	I/O	these terminals. Data I/O. The Dn signals are bidirectional and pass data between the PHY and the LLC. Bus-keeper circuitry is built into these terminals.
12, 13	11/2		

^{*} Active-low signals are indicated by "/" at the beginning of signal names, within this document.

Table 1. Signal Descriptions (continued)

Pin	Signal*	Туре	Name/Description
23	/ISO	I	Link Interface Isolation Disable Input (Active-Low). /ISO centrols the operation of an internal pulse differentiating function used on the PHY-LLC interface signals, CTLn and Dn, when they operate as outputs. When /ISO is asserted low, the isolation barrier is implemented between PHY and its LLC (as described in Annex J of IEEE 1394-1995). /ISO is normally tied high to disable isolation differentiation. Bus-keepers are enabled when /ISO is high (inactive) on CTLn, Dn, and LREQ. When /ISO is low (active), the bus-keepers are disabled. Please refer to Agere's application note IEEE 1394 Isolation (AP98-074 CMPR) for more information.
16	LPS	I	Link Power Status. LPS is connected to either the VDD supplying the LLC or to a pulsed output that is active when the LLC is powered for the purpose of monitoring the LLC power status. If LPS is inactive for more than 1.2 μ s and less than 25 μ s, the PHY-link interface is reset. If LPS is inactive for greater than 25 μ s, the PHY will disable the PHY/link interface to save power. FW803 continues its repeater function.
1	LREQ	ļ	Link Request. LREQ is an output from the LLC that requests the PHY to perform some service. Bus-keeper circuitry is built into this terminal.
20	PC0	I	Power-Class Indicators. On hardware reset (/RESET), these inputs set the
21	PC1		default value of the power class indicated during SelfID. These bits can be tied to VDD (high) or to ground (low) as required for particular power consump-
22	PC2		tion and source characteristics. In SelfID packet (see Section 4.3.4.1 of the 1394a-2000 Specification), PC0, the most significant bit of this 3-bit field, corresponds to bit 20, PC1 corresponds to bit 21, and PC2 corresponds to bit 22. As an example, for a Power_Class value of 001, PC0 = 0, PC1 = 0, and PC2 = 1.
19	PD	I	Powerdown. When asserted high, PD turns off all internal circuitry except the bias-detect circuits that drive the CNA signal. Internal FW803 logic is kept in the reset state as long as PD is asserted. The PD terminal is provided for backward-compatibility.
57	VDDPLL	_	Power for PLL Circuit. VDDPLL supplies power to the PLL circuitry portion of the device.
58	VssPLL		Ground for PLL Circuit. VSSPLL is tied to a low-impedance ground plane.
54	R0		Current Setting Resistor. An internal reference voltage is applied to a resistor connected between R0 and R1 to set the operating current and the cable driver output current. A low temperature-coefficient resistor (TCR) with
55	R1		a value of 2.49 k Ω \pm 1% should be used to meet the IEEE 1394-1995 standard requirements for output voltage limits.
61	/RESET		Reset (Active-Low). When /RESET is asserted low (active), a 1394 bus reset condition is set on the active cable port and the FW803 is reset to the reset start state. To guarantee that the PHY will reset, this pin must be held low for at least 2 ms. An internal pull-up resistor connected to VDD is provided so that only an external delay capacitor $(0.1~\mu F)$ and resistor $(510~k\Omega)$ in parallel, are required to connect this pin to ground. This circuitry will ensure that the capacitor will be discharged when PHY power is removed. The input is standard logic buffer and can also be driven by an open-drain logic output buffer. Do not leave this pin unconnected.

^{*} Active-low signals are indicated by "/" at the beginning of signal names, within this document.

Table 1. Signal Descriptions (continued)

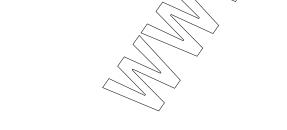
Pin	Signal*	Туре	Name/Description
28	SE	I	Test Mode Control. SE is used during Agere's manufacturing test and should be tied to Vss for normal operation.
29	SM	_	Test Mode Control. SM is used during Agere's manufacturing test and should be tied to Vss for normal operation.
63	SYSCLK	0	System Clock. SYSCLK provides a 49.152 MHz clock signal, which is synchronized with the data transfers to the LLC.
36	TPA0+	Analog I/O	Port0, Port Cable Pair A. TPA0± is the port A connection to the twisted-pair
35	TPA0-		cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.
41	TPA1+	Analog I/O	Port1, Port Cable Pair A. TPA1± is the port A connection to the twisted-pair
40	TPA1-		cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.
47	TPA2+	Analog I/O	Port2, Port Cable Pair A. TPA2± is the port A connection to the twisted-pair
46	TPA2-		cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.
34	TPB0+	Analog I/O	Port0, Port Cable Pair B. TPB0± is the port B connection to the twisted-pair
33	TPB0-		cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.
39	TPB1+	Analog I/O	Port1, Port Cable Pair B. TPB1± is the port B connection to the twisted-pair
38	TPB1-		cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.
45	TPB2+	Analog I/O	Port2, Port Cable Pair B. TPB2± is the port B connection to the twisted-pair
44	TPB2=	\ \ \	cable. Board traces from each pair of positive and negative differential signal pins should be kept as short as possible and matched to the external load resistors and to the cable connector. When the FW803's 1394 port pins are not wired to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.

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Table 1. Signal Descriptions (continued)

Pin	Signal*	Туре	Name/Description						
37	TPBIAS0	Analog I/O	Portn, Twisted-Pair Bias. (where n is the port number) TPBIAS provides						
42	TPBIAS1		the 1.86 V nominal bias voltage needed for proper operation of the twisted- pair cable drivers and receivers and for sending a valid cable connection						
48	TPBIAS2		signal to the remote nodes. When the FW803's 1394 port pins are not we to a connector, the unused port pins may be left unconnected. Internal connect-detect circuitry will keep the port in a disconnected state.						
7, 17, 26, 27, 62	VDD	_	Digital Power. VDD supplies power to the digital portion of the device.						
30, 31, 43, 50, 51	Vdda	_	Analog Circuit Power. VDDA supplies power to the analog portion of the device.						
2, 14, 25, 56, 64	Vss	_	Digital Ground. All Vss signals should be tied to the low-impedance ground plane.						
32, 49, 52, 53	Vssa	_	Analog Circuit Ground. All VSSA signals should be tied together to a low-impedance ground plane.						
59	ΧI	_	Crystal Oscillator. XI and XO connect to a 24.576 MHz parallel resonant fundamental mode crystal. Although, when a 24.576 MHz clock source is used, it can be connected to XI with XO left unconnected. The optimum						
60	ХО		values for the external load capacitors and resistor are dependent on the specifications of the crystal used. It is necessary to add an external series resistor (RL) to the XO pin (see Figures 3 and 5). For more details, refer to the Crystal Selection Considerations section in this data sheet. Note that it is very important to place the crystal as close as possible to the XO and XI pins, i.e., within 0.5 in X1.27 cm.						

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Application Information

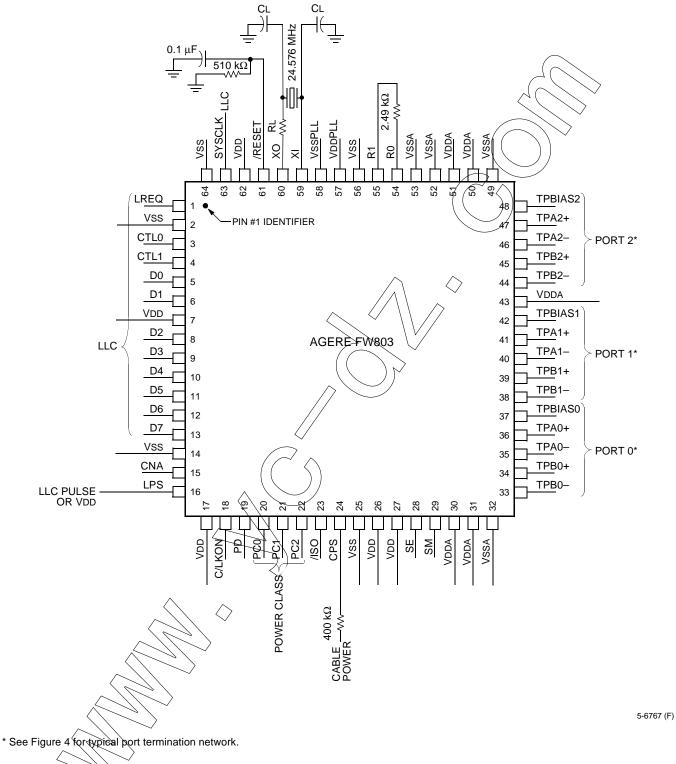


Figure 3. Typical External Component Connections

Application Information (continued)

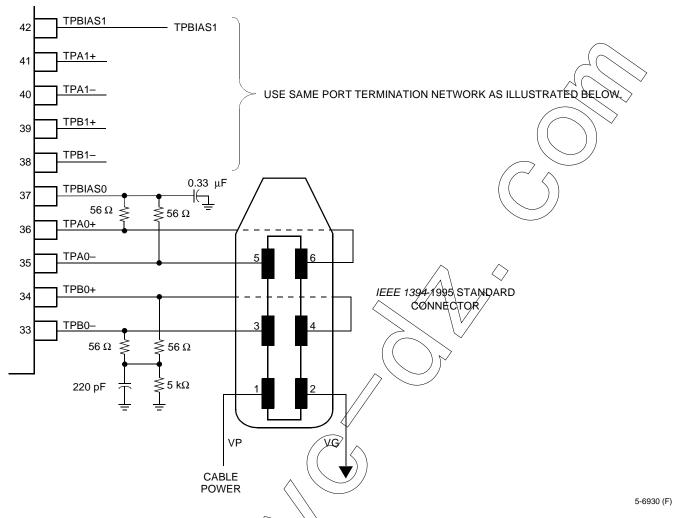


Figure 4. Typical Port Termination Network

Crystal Selection Considerations

The FW803 is designed to use an external 24:576 MHz parallel resonant fundamental mode crystal connected between the XI and XO terminals to provide the reference for an internal oscillator circuit. The *IEEE 1394*a-2000 standard requires that FW803 have less than ±100 ppm total variation from the nominal data rate, which is directly influenced by the crystal. To achieve this, it is recommended that an oscillator with a nominal 50 ppm or less frequency tolerance be used.

The total frequency variation must be kept below ± 100 ppm from nominal with some allowance for error introduced by board and device variations. Trade-offs between frequency tolerance and stability may be made as long as the total frequency variation is less than ± 100 ppm.

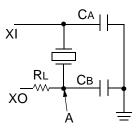
Crystal Selection Considerations (continued)

Load Capacitance

The frequency of oscillation is dependent upon the load capacitance specified for the crystal, in parallel resonant mode crystal circuits. Total load capacitance (CL) is a function of not only the discrete load capacitors, but also capacitances from the FW803 board traces and capacitances of the other FW803 connected components.

The values for load capacitors (CA and CB) should be calculated using this formula:

 $CA = CB = (CL - Cstray) \times 2$



Where:

CL = load capacitance specified by the crystal manufacturer
Cstray = capacitance of the board and the FW803, typically 2pF—3 pF

RL = load resistance; the value of RL is dependent on the specific crystal used. Rease refer to your crystal manufacturer's data sheet and application notes to determine an appropriate value.

Figure 5. Crystal Circuitry

Adjustment to Crystal Loading

The resistor (RL) in Figure 5 is recommended for fine tuning the crystal circuit. The value for this resistor is dependent on the specific crystal used. Please refer to your crystal manufacturer's data sheet and application notes to determine an appropriate value for RL. A more precise value for this resistor can be obtained by placing different values of RL on a production board and using an oscilloscope to view the resultant clock waveform at node A for each resistor value. The desired waveform should have the following characteristics: the waveform should be sinusoidal, with an amplitude as large as possible, but not greater than 3.3 V or less than 0 volts.

Crystal/Board Layout

The layout of the crystal portion of the PHY circuit is important for obtaining the correct frequency and minimizing noise introduced into the FW803 PLL. The crystal and two load capacitors (CA + CB) should be considered as a unit during layout. They should be placed as close as possible to one another, while minimizing the loop area created by the combination of the three components. Minimizing the loop area minimizes the effect of the resonant current that flows in this resonant circuit. This layout unit (crystal and load capacitors) should then be placed as close as possible to the PHY XL and XO terminals to minimize trace lengths. Vias should not be used to route the XI and XO signals.

1394 Application Support Contact Information

E-mail: support1394@agere.com

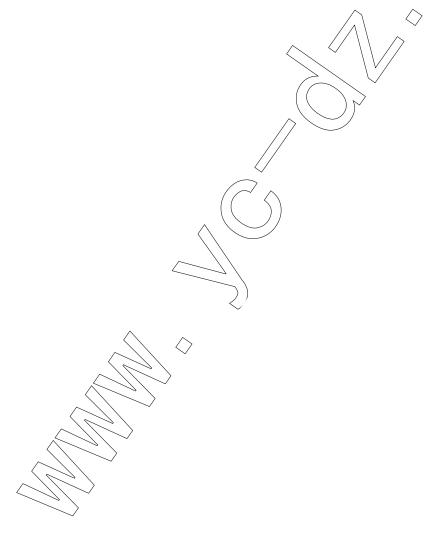
Absolute Maximum Ratings

Stresses in excess of the absolute maximum ratings can cause permanent damage to the device. These are absolute stress ratings only. Functional operation of the device is not implied at these or any other conditions in excess of those given in the operational sections of the data sheet. Exposure to absolute maximum ratings for extended periods can adversely affect device reliability.

Table 2. Absolute Maximum Ratings

Parameter	Symbol	Min	Max	Unit
Supply Voltage Range	VDD	3.0	3.6	V
Input Voltage Range*	Vı	-0.5	VDD + 0.5	V
Output Voltage Range at Any Output	Vo	-0.5	(VDD + 0.5	V
Operating Free Air Temperature	TA	0	70	°C
Storage Temperature Range	Tstg	– 65	150	°C

^{*} Except for 5 V tolerant I/O (CTL0, CTL1, D0—D7, and LREQ) where VI max = 5.5 V.



Electrical Characteristics

Table 3. Analog Characteristics

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Supply Voltage	Source power node	VDD-SP	3.0 (3.3	3.6	V
Differential Input Voltage	Cable inputs, 100 Mbits/s operation	VID—100	142		²⁶⁰	mV
	Cable inputs, 200 Mbits/s operation	VID-200	132	\rightarrow	260	mV
	Cable inputs, 400 Mbits/s operation	VID-400	100	\rightarrow	260	mV
	Cable inputs, during arbitration	VID—ARB	168		265	mV
Common-mode Voltage	TPB cable inputs,	VcM	1.165	_	2.515	V
Source Power Mode	speed signaling off					
	TPB cable inputs,	VCM—(SP—100)	1.165	_	2.515	V
	S100 speed signaling on TPB cable inputs,	VCM—SP—200	0.935		2.515	V
	S200 speed signaling on	VCM—SP—200	0.935	_	2.515	V
	TPB cable inputs,	VCM—SP—400	0.532		2.515	V
	S400 speed signaling on		0.002		2.010	•
Common-mode Voltage	TPB cable inputs,	√VCM	1.165		2.015	V
Nonsource Power Mode*	speed signaling off	\ //				
	TPB cable inputs,	VCM—NSP—100	1.165	_	2.015	V
	S100 speed signaling on	~				
	TPB cable inputs,	VCM—NSP—200	0.935	_	2.015	V
	S200 speed signaling on		0.500		0.045	
	TPB cable inputs,	VCM—NSP—400	0.532	_	2.015	V
Receive Input Jitter	S400 speed signaling on TPA, TPB cable inputs,				1.08	ns
Treceive input sitter	100 Mb/ts/s operation	_			1.00	113
	TPA, TPB cable inputs,	_			0.5	ns
	200 Mbits/s operation					
	TPA, TPB cable inputs,	_	_		0.315	ns
	400 Mbits/s operation					
Receive Input Skew	Between TPA and TPB cable inputs,	_	_	_	0.8	ns
	100 Mbits/s operation					
	Between TPA and TPB cable inputs,	_	_	_	0.55	ns
	200 Mbits/s operation Between TPA and TPB cable inputs,				0.5	no
_	400 Mbits/s operation	_	_		0.5	ns
Positive Arbitration	—	VTH+	89		168	mV
Comparator Input						
Threshold Voltage						
Negative Arbitration	<u> </u>	VTH-	-168	_	-89	mV
Comparator Input						
Threshold Voltage	000 Mb %s /s	\/=	45		400	\/
Speed Signal Input Threshold Voltage	200 Mbits/s	VTH—S200	45		139	mV
Output Current	400 Mbits/s TPBIAS outputs	VTH—S400	266 -5	_	445 2.5	mV m^
TPBIAS Output Voltage	At rated I/O current	Vo	_5 1.665		2.015	mA V
Current Source for	At Taleu I/O Current	ICD	1.003		76	
Connect Detect Circuit		ICD			70	μΑ
				l		<u> </u>

^{*} For a node that does not source power (see Section 4.2.2.2 in *IEEE 1394*-1995 Standard).

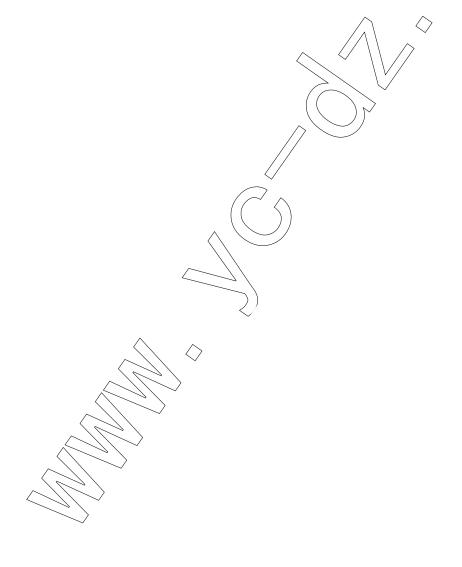
Electrical Characteristics (continued)

Table 4. Driver Characteristics

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Differential Output Voltage	56 Ω load	Vod	172	— <i>,</i>	265	m۷
Off-state Common-mode Voltage	Drivers disabled	Voff	_	-4	20	mV
Driver Differential Current, TPA+, TPA-, TPB+, TPB-	Driver enabled, speed signaling off*	IDIFF	-1.05		1.05	mA
Common-mode Speed Signaling Current, TPB+, TPB-	200 Mbits/s speed signaling enabled [†]	ISP	-2.53		4.84	mA
	400 Mbits/s speed signaling enabled [†]	ISP	-8.1	$\widehat{}$	-12.4	mA

^{*} Limits are defined as the algebraic sum of TPA+ and TPA- driver currents. Limits also apply to TPB+ and TPB_as the algebraic sum of driver currents.

[†] Limits are defined as the absolute limit of each of TPB+ and TPB- driver currents.



Electrical Characteristics (continued)

Table 5. Device Characteristics

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Supply Current: One Port Active All Ports Active No Ports Active, LPS = 0 PD = 1	VDD = 3.3 V	IDD IDD IDD IDD	_ _ _	113 151 85	\\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	mA mA mA
High-level Output Voltage	Iон max, VDD = min	Voн	VDD - 0.4	ر کے ا	<u> </u>	V
Low-level Output Voltage	IOL min, VDD = max	Vol			0.4	V
High-level Input Voltage	CMOS inputs	ViH	0(7(VDD)	_	_	V
Low-level Input Voltage	CMOS inputs	VIL		_	0.2 VDD	V
Pull-up Current, /RESET Input	VI = 0 V	lı	11	1	32	μΑ
Powerup Reset Time, /RESET Input	VI = 0 V		⊘ 2		_	ms
Rising Input Threshold Voltage /RESET Input	- <	WRST	1.1		1.4	V
Output Current	SYSCLK	OL/IOH	-16	_	16	mA
	Control, data	IOL/IOH @ CMOS	-12	_	12	mA
	CNA//	loL/loн	-16	_	16	mA
	C/LKON	loL/loн	-2	_	2	mA
Input Current, LREQ, LPS, PD, SE, SM, PC[0:2] Inputs	VI = VDD or 0 V	lı	_	_	°±1	μА
Off-state Output Current, CTL[0:1], D[0:7], C/LKON I/Os	VO = VDD or 0 V	loz	_	_	°±5	μΑ
Power Status Input Threshold Voltage, CPS Input	400 kΩ resistor	VTH	7.5	_	8.5	V
Rising Input Threshold Voltage*, LREQ, CTLn, Dn	_	VIT+	VDD/2 + 0.3	_	VDD/2 + 0.8	V
Falling Input Threshold Voltage*, LREQ, CTLn, Dn	_	VIT-	VDD/2 - 0.8	_	VDD/2 - 0.3	V
Bus Holding Current, LREQ, CTLn, Dn	VI = 1/2(VDD)	_	250	_	550	μΑ
Rising Input Threshold Voltage LPS	_	VLIH	_	_	0.24 VDD + 1	V
Falling Input Threshold Voltage LPS	_	VLIL	0.24 VDD + 0.2	_	_	V

^{*} Device is capable of both differentiated and undifferentiated operation.

Timing Characteristics

Table 6. Switching Characteristics

Symbol	Parameter	Measured	Test Conditions	Min	Тур	Max	Unit
_	Jitter, Transmit	TPA, TPB	_	_		0.15	ns
_	Transmit Skew	Between TPA and TPB	_	_	4/	±0.1	ns
tr	Rise Time, Transmit (TPA/TPB)	10% to 90%	Rι = 56 Ω, Cι = 10 pF	-(1.2	ns
tf	Fall Time, Transmit (TPA/TPB)	90% to 10%	RI = 56 Ω , CI = 10 pF)	1.2	ns
tsu	Setup Time, Dn, CTLn, LREQ↑↓ to SYSCLK↑	50% to 50%	See Figure 6.	6) —	_	ns
th	Hold Time, Dn, CTLn, LREQ↑↓ from SYSCLK↑	50% to 50%	See Figure 6.	0		_	ns
td	Delay Time, SYSCLK↑ to Dn, CTLn↑↓	50% to 50%	See Figure 7.	1	_	6	ns

Table 7. Clock Characteristics

Parameter	Symbol	Min	Тұ	(b)	Max	Unit
External Clock Source Frequency	f	24.5735	24,5	760	24.5785	MHz

Timing Waveforms

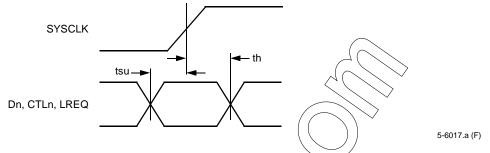


Figure 6. Dn, CTLn, and LREQ Input Setup and Hold Times Waveforms

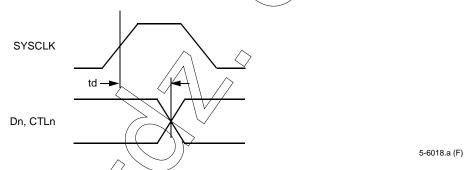
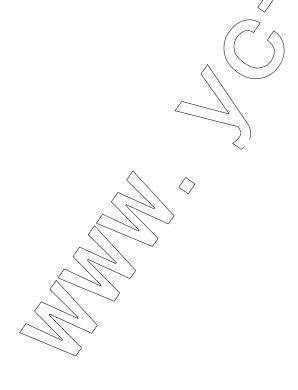


Figure 7. Dn, CTLn Output Delay Relative to SYSCLK Waveforms



Internal Register Configuration

The PHY register map is shown below in Table 8. (Refer to IEEE P1394a Draft 2.0, Section 6.1 for more information).

Table 8. PHY Register Map

Address				Con	tents				
	Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7	
00002			Physic	al_ID			R	→ PS	
00012	RHB	IBR			Gap	o_count			
00102	E	extended (7)		XXXXX		Tota	l_ports		
00112	1	Max_speed		XXXXX		(p	elay		
01002	Link_active	Contender		Jitter			Pwr_class		
01012	Resume_int	ISBR	Loop	Pwr_fail	Timeout	Port_event	Enab_accel	Enab_multi	
01102	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	
01112	F	Page_select				Port	_select		
10002				Register 0	Page_select				
:	•					V		:	
11112		Register 7 Page_select							
		REQUIRED		XXXXX	RESERVE)			
		•							

The meaning of the register fields within the PHY register map are defined by Table 9 below. Power reset values not specified are resolved by the operation of the PHY state machines subsequent to a power reset.

Table 9. PHY Register Fields

Field	Size	Туре	Power Reset Value	Description
Physical_ID	6	r	000000	The address of this node determined during self-identification. A value of 63 indicates a malconfigured bus; the link will not transmit any packets.
R	1	r	√0	When set to one, indicates that this node is the root.
PS	1	r		Cable power active.
RHB	1	rw		Root Hold-off Bit. When set to one, the force_root variable is TRUE, which instructs the PHY to attempt to become the root during the next tree identify process.
IBR	1 <		0	Initiate Bus Reset. When set to one, instructs the PHY to set ibr TRUE and reset_time to RESET_TIME. These values in turn cause the PHY to initiate a bus reset without arbitration; the reset signal is asserted for 166 μs. This bit is self-clearing.
Gap_count	6	*	3F16	Used to configure the arbitration timer setting in order to optimize gap times according to the topology of the bus. See Section 4.3.6 of <i>IEEE</i> Standard <i>1394</i> -1995 for the encoding of this field.
Extended	3	r	7	This field has a constant value of seven, which indicates the extended PHY register map.

Table 9. PHY Register Fields (continued)

Field	Size	Туре	Power Reset Value	Description				
Total_ports	4	r	3	The number of ports implemented by this RHY. This count reflects the number.				
Max_speed	3	r	0102	Indicates the speed(s) this PHY supports: 0002 = 98.304 Mbits/s 0012 = 98.304 and 196.608 Mbits/s 0102 = 98.304, 196.608, and 393.216 Mbits/s 0112 = 98.304, 196.608, 393.216, and 786.43 Mbits/s 1002 = 98.304, 196.608, 393.216, 786.432, and 1,572.864 Mbits/s 1012 = 98.304, 196.608, 393.216, 786.432, 1,572.864, and 3,145.728 Mbits/s All other values are reserved for future definition.				
Delay	4	r	0000	Worst-case repeater delay, expressed as 144 + (delay * 20) ns.				
Link_active	1	rw	See description.	Link Active. Cleared or set by software to control the value of the L bit transmitted in the node's self-ID packet 0, which will be the logical AND of this bit and LPS active.				
Contender	1	rw	See description.	Cleared or set by software to control the value of the C bit transmitted in the self-ID packet. Powerup reset value is set by C/LKON pin.				
Jitter	3	r	000	The difference between the fastest and slowest repeater data delay, expressed as (jitter + 1) * 20 ns.				
Pwr_class	3	rw	See description.	Power-Class. Controls the value of the pwr field transmitted in the self-ID packet. See Section 4.3.4.1 of <i>IEEE</i> Standard 1394-1995 for the encoding of this field. PC0, PC1, and PC2 pins set up power reset value.				
Resume_int	1	rw	0	When set to one, the PHY will set Port_event to one if resume operations commence for any port.				
ISBR	1	rw	0	Initiate Short (Arbitrated) Bus Reset. A write of one to this bit instructs the PHY to set ISBR true and reset_time to SHORT_RESET_TIME. These values in turn cause the PHY to arbitrate and issue a short bus reset. This bit is self-clearing.				
Loop	1	rw	Ø	Loop Detect. A write of one to this bit clears it to zero.				
Pwr_fail	1	rw	1	Cable Power Failure Detect. Set to one when the PS bit changes from one to zero. A write of one to this bit clears it to zero.				
Timeout	1	TA	0	Arbitration State Machine Timeout. A write of one to this bit clears it to zero (see MAX_ARB_STATE_TIME).				
Port_event		rw >	0	Port Event Detect. The PHY sets this bit to one if any of connected, bias, disabled, or fault change for a port whose Int_enable bit is one. The PHY also sets this bit to one if resume operations commence for any port and Resume_int is one. A write of one to this bit clears it to zero.				

Table 9. PHY Register Fields (continued)

Field	Size	Туре	Power Reset Value	Description				
Enab_accel	1	rw	0	Enable Arbitration Acceleration. When set to one, the PHY will use the enhancements specified in clause 7.10. PHY behavior is unspecified if the value of Enab_accel is changed while a bus request is pending.				
Enab_multi	1	rw	0	Enable Multispeed Packet Concatenation. When set to one, the link will signal the speed of all packets to the PHY.				
Page_select	3	rw	000	Selects which of eight possible PHY register pages are accessible through the window at PHY register addresses 10002 through 11112, inclusive.				
Port_select	4	rw	000	If the page selected by Page_select presents per-port information, this field selects which port's registers are accessible through the window at PHY register addresses 10002 through 11112, inclusive. Ports are numbered monotonically starting at zero, p0.				

The port status page is used to access configuration and status information for each of the PHY's ports. The port is selected by writing zero to Page_select and the desired port number to Port_select in the PHY register at address 01112. The format of the port status page is illustrated by Table 10 below, reserved fields are shown shaded. The meanings of the register fields with the port status page are defined by Table 11.

Table 10. PHY Register Page 0: Port Status Page

Address	Contents
Address	_ Contents

	Bit 0	Bit 1	Bit 2	∫ (Bit 3 < \ \	Bit 4	Bit 5	Bit 6	Bit 7
10002	ASt	at	BStat		Child	Connected	Bias	Disabled
10012	Neg	otiated_spe	ed	Int_enable	Fault	XXXXX	XXXXX	XXXXX
10102	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX
10112	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX
11002	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX
11012	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX
11102	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX
11112	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX

REQUIRED XXXXX RESERVED

The meaning of the register fields with the port status page are defined by Table 11 below.

Table 11. PHY Register Port Status Page Fields

	Ι			
Field	Size	Туре	Power Reset Value	Description
AStat	2	r		TPA line state for the port: 002 = invalid 012 = 1 102 = 0 112 = Z
BStat	2	r	_	TPB line state for the port (same encoding as AStat).
Child	1	r	0	If equal to one, the port is a child; otherwise, a parent. The meaning of this bit is undefined from the time a bus reset is detected until the PHY transitions to state T1: Child Handshake during the tree identify process (see Section 4.4.2.2 in IEEE Standard 1394-1995).
Connected	1	r	0	If equal to one, the port is connected.
Bias	1	r	0	If equal to one, incoming TPBIAS is detected.
Disabled	1	rw	0	If equal to one, the port is disabled.
Negotiated_speed	3	r	000	Indicates the maximum speed negotiated between this PHY port and its immediately connected port; the encoding is the same as for the PHY register Max_speed field.
Int_enable	1	rw	0	Enable Port Event Interrupts. When set to one, the PHY will set Port_event to one if any of connected, bias, disabled, or fault (for this port) change state.
Fault	1	rw	0	Set to one if an error is detected during a suspend or resume operation. A write of one to this bit clears it to zero.



The vendor identification page is used to identify the PHY's vendor and compliance level. The page is selected by writing one to Page_select in the PHY register at address 01112. The format of the vendor identification page is shown in Table 12; reserved fields are shown shaded.

Table 12. PHY Register Page 1: Vendor Identification Page

Address	Contents								
	Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7	
10002		Compliance_level							
10012	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	XXXXX	
10102									
10112				Vend	or_ID				
11002									
11012									
11102				Produ	ıct_ID				
11112									
		_			_				
		REQUIRED	•	XXXXX	RESERVE)			

The meaning of the register fields within the vendor identification page are defined by Table 13.

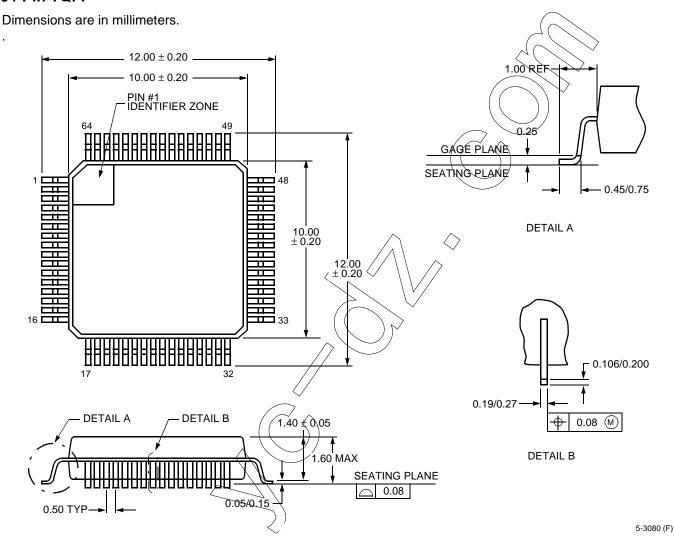
Table 13. PHY Register Vendor Identification Page Fields

Field	Size	Туре	Description
Compliance_level	8	r	Standard to which the PHY implementation complies: 0 = not specified 1 = IEEE 1394a-2000 Agere's FW803 compliance level is 1. All other values reserved for future standardization.
Vendor_ID	24	r	The company ID or organizationally unique identifier (OUI) of the manufacturer of the PHY. Agere's vendor ID is 00601D16. This number is obtained from the <i>IEEE</i> registration authority committee (RAC). The most significant byte of Vendor_ID appears at PHY register location 10102 and the least significant at 11002.
Product_ID	24	r	The meaning of this number is determined by the company or organization that has been granted Vendor_ID. Agere's FW803 product ID is 08140216. The most significant byte of Product_ID appears at PHY register location 11012 and the least significant at 11112.

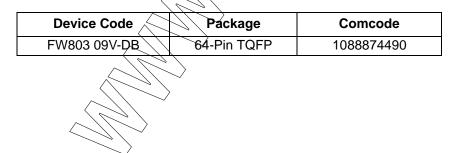
The vendor-dependent page provides access to information used in manufacturing test of the FW803.

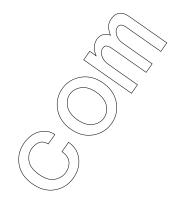
Outline Diagrams

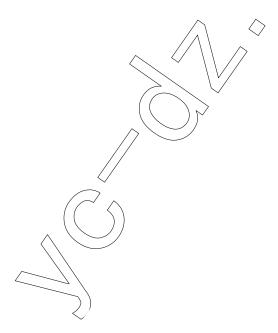
64-Pin TQFP



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